

A Novel Dual Inductor DC-DC Buck Converter

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Abstract: A novel DC-DC Buck Converter using dual inductor and switched capacitor is proposed. Proposed circuit consists of two inductor instead of single inductor used in conventional dc-dc buck converter. The proposed converter offers better efficiency in obtaining same low voltage gain. Energy is effectively transferred from source to load using one inductor and capacitor while other inductor is used to step down the voltage and also get better efficiency than conventional dc-dc buck converter. In switch on time period the two inductors are in parallel and in off time period the two inductors are in series, on that time two inductors charged the output capacitor to get better efficiency. Simulation results verified the high efficiency three-stage operation of the proposed converter.

Keywords: Dual inductor, Switched capacitor, DC-DC Buck Converter

I. Introduction

Power converter plays a very major role in wide range of applications of power electronics. SMPS (Switched Mode Power Supply) converters are very important in sense of ease of application & scope of improvement. A dc-dc buck converter is a power converter which is step down the voltage, while stepping up the current from its input to output. It is a class of SMPS (Switched Mode Power Supply) typically containing at least a diode and a transistor, although modern buck converters frequently replace the diode with a second transistor used for synchronous rectification and at least one energy storage element, a capacitor, inductor, or the two in combination [1, 2]. To reduce voltage ripple, filters made of capacitors are normally added to such a converter's output and input. DC-DC converters provide much greater power efficiency than linear regulators, which are simpler circuits that lower voltages by dissipating power as heat, which does not step up output current [3, 4].

The basic operation of the buck converter has the current in an inductor controlled by two switches, usually a transistor and a diode. In the idealized converter, all the components are considered to be perfect. Specifically, the switch and the diode have zero voltage drop when on and zero current flow when off and the inductor has zero series resistance. Further, it is assumed that the input and output voltages do not change over the course of a cycle. This would imply the output capacitance as being infinite. The buck converter is best understood in terms of the relation between current and voltage of the inductor. Beginning with the switch open, the current in the circuit is zero. When the switch is first closed, the current will begin to increase, and the inductor will produce an opposing voltage across its terminals in response to the changing current. This voltage drop counteracts the voltage of the source and therefore reduces the net voltage across the load. Over time, the rate of change of current decreases, and the voltage across the inductor also then decreases, increasing the voltage at the load.

Buck converter aims at efficiencies higher than a linear regulator for a large range of the output current [5]. If the generated voltage V_{out} is D times the supply voltage, the efficiency of the linear regulator is D . Therefore, the goal is to keep system efficiency well above D for an extended range of output current. Just obtaining very high switching frequency [6, 7] is not practical. The goal of this design is to achieve for I-A output current, efficiency higher than what reported in published research results [8, 9].

The proposed DC-DC Buck Converter comprises a DC input voltage, active power switch, dual inductor, three diodes, one output capacitor respectively. This proposed converters use a pair of switches, usually one controlled (e.g. MOSFET) and one uncontrolled (i.e. diode), to achieve unidirectional power flow from input to output. The converters also use one capacitor and one inductor to store and transfer energy from input to output. Another inductor uses to get better efficiency than conventional dc-dc buck converter.

DC-DC converters can be operated either in continuous conduction mode (CCM) or in discontinuous conduction mode (DCM). DC-DC converters that operated in DCM provide faster transient response (due to its low inductance) at the expense of higher device stresses. A fixed frequency PWM based sliding mode controllers for dc-dc converters operating in DCM [10, 11]. Buck converter when operated in CCM, gives a continuous output current, with smaller current ripple and low switching noise. CCM operation is usually preferred for large

current applications, because it can deliver more current than the converter operating in DCM. However, a DCM converter has a much faster transient response and a loop gain that is easier to compensate than a CCM converter. Hence, for fulfill of both the requirements, a new converter that combines the advantage of both CCM and DCM converters is developed. CCM buck converter has much improved current handling capability with reduced current and voltage ripple [12]. For this proposed circuit only consider the DC-DC Buck Converters operated in CCM.

II. Operating Principle

In DC-DC Converters, the average dc output voltage must be controlled to equal a desired level, though the input voltage and the output load fluctuate. Switch-mode dc-dc converters utilize one or more switches to transform dc from one level to another. In dc-dc converters with a given input voltage, the average output voltage is controlled by controlling the switch on and off duration. The average value of the output voltage depends on the switch on and off duration. Basic DC-DC Buck Converter comprises a DC input voltage, active power switch, one inductor, one diode and one capacitor respectively. Buck converters can be remarkably efficient, higher than 90%.

The circuit configuration of the proposed DC-DC Buck Converter is shown in Fig. 1. It comprises a DC input voltage (V_{in}), active power switch (S), dual inductor ($L1$ & $L2$), three diodes ($D1$, $D2$ & $D3$), capacitor (CL) respectively. The capacitor (CL) is employed as an output capacitor with the value of $200 \mu F$ which store the energy. Inductor ($L1$) with the value of $66 mH$, employed to store and transfer energy from input to output. They also filters or smooth voltage and current. Another inductor ($L2$) with the value of $66 mH$ is used to get better efficiency than conventional dc-dc buck converter.

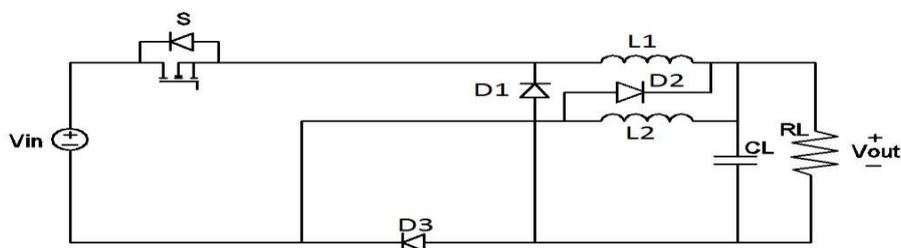


Fig. 1. Circuit configuration of proposed converter

In order to simplify the circuit analysis of the converter, some assumptions are considered as follow:

1. The capacitor is sufficiently large. Therefore V_C and V_{out} are considered being constant during one switching period.
2. All components are ideal.

According to the assumptions, the CCM operation of the proposed converter includes three intervals one switching period. The current flow path of the proposed converter for each stage is depicted in Fig. 2. The operating stages are explained as follows:

- 1) **Stage I:** In this stage the switch is turned on. Also diode $D3$ is turned on and diode $D1$ & $D2$ are turned off. The DC source ($V1$) for a time duration DT , the switch (S) conducts inductor ($L1$) current, diode ($D3$) current and another two diodes ($D1$ & $D2$) becomes reverse biased. The required energy of load (RL) is supplied by the output capacitor (CL). Again the DC source ($V1$) for time duration DT , the switch (S) conducts inductors ($L1$ & $L2$) currents and another two diodes ($D1$ & $D2$) becomes reverse biased, shown in Fig. 2(a). This result is in positive voltage V_L across the inductor ($L2$). This voltage causes a linear increase in the inductor current i_L .

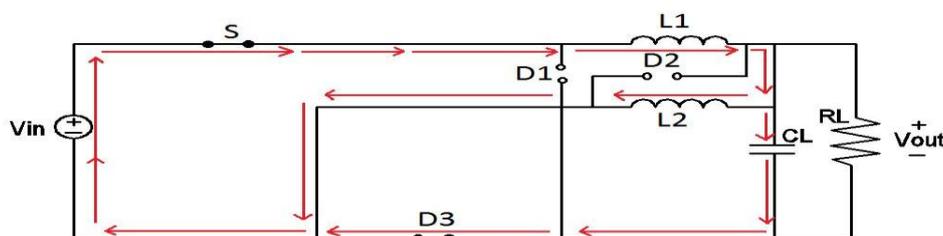


Fig 2(a). Switch on period

- 2) **Stage II:** In this stage the switch is turned off. Also diode $D2$ is turned off and diode $D1$ is turned on because of the inductive energy storage, $iL1 = iL$ continuous to flow. This current now flows through the capacitor (CL) and diode ($D1$). There is no current flow in the another inductor ($L2$), shown in Fig. 2(b) for a time duration $(1-D)T$ until the switch is turned on again.

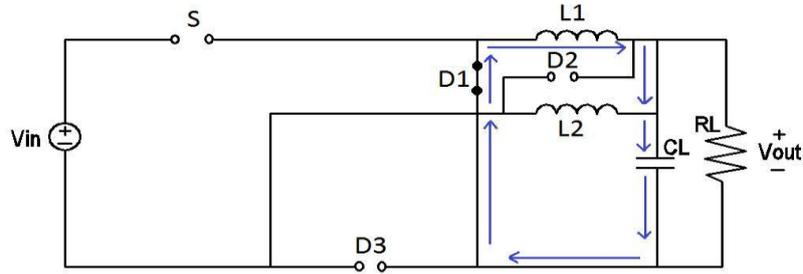


Fig.2(b). Switch off period

III. Steady-State Operation Of The Proposed Converter

To simplify the steady state analysis during the time when switch is on, the following equation can be written according to Fig. 2(a):

$$V_{in} - VL1 - V_{out} = 0$$

$$V_{in} - V_{out} = VL1 \text{-----(i)}$$

Now, from the following equation can be written according to Fig. 2(b):

$$V_{in} - VL1 - VL2 = 0$$

$$V_{in} = VL1 + VL2$$

$$V_{in} = 2VL1 \text{----- (ii) [VL1 = VL2]}$$

From the Fig. 1, it is known that the values of inductor $L1$ and $L2$ are same. So, we can say from there, that –

$$VL1 = VL2$$

Now adding equation (i) and (ii), we can get –

$$VL1 = (2V_{in} - V_{out})/3$$

This result is in positive voltage, $VL = VL1 = (2V_{in} - V_{out})/3$

When switch is off, the following equation can be written according to Fig.2(c):

$$-VL1 - V_{out} = 0$$

$VL1 = VL = -V_{out}$ zero yields to the

Equating the integral of inductor voltage over one time period

$$\int_0^T V_L dt = \int_0^{t_{on}} V_1 dt + \int_0^{t_{off}} V_L dt = 0$$

$$\frac{\{(2V_{in} - V_{out}) * DT\}}{3} + \{(-V_{out}) * (1 - D)T\} = 0$$

$$\frac{2V_{in}DT - V_{out}DT}{3} - V_{out}T + V_{out}DT = 0$$

$$\frac{2V_{in}DT - V_{out}DT - 3V_{out}T + 3V_{out}DT}{3} = 0$$

$$2V_{in}DT - 2V_{out}DT - 3V_{out}T = 0$$

$$2V_{in}DT - 2V_{out}DT = 3V_{out}T$$

$$2V_{in}D - 2V_{out}D = 3V_{out}$$

$$2V_{in}D = 3V_{out} + 2V_{out}D$$

$$2V_{in}D = V_{out}(3 + 2D)$$

$$G = \frac{2D}{3 + 2D}$$

The voltage gain of this proposed circuit is-

IV. Performance Of The Proposed Circuit

The simulation of the proposed circuit is performed using PSIM and the results obtained are presented in Table 2. The source is a DC source with input voltage 1200V because of to take better efficiency than conventional. The switching frequency of the power switch has been set to 8KHz and an MOSFET Switch is considered as the switching device. The circuit is simulated with the component $L1$ & $L2$ as 66mH, output filter capacitor CL as 200μF and load resistance RL as 50Ω. The proposed circuit has been compared with the conventional DC-DC basic buck using same circuit parameter values. The specifications of the implemented circuit are also given in Table 1. Typical results are presented in the next subsection.

Table 1 Specifications Of The Implemented Parameters

Specifications	Values
Input DC Voltage	$V_{in} = 1200 V$
Output DC Voltage	$V_{out} = 190 V$
Inductor	$L1 = L2 = 66 mH$
Capacitor	$CL = 200 \mu F$
Resistor	$RL = 50 \Omega$

A. Result from Simulation:

The typical wave shapes of input voltage and output voltage along with the spectrum of the input current for the proposed Buck based input side switched DC-DC converter is shown in Fig. 3. It is evident from the wave shapes obtained that the output voltage of the proposed Buck based DC-DC converter is almost DC signal.

The experimental results are shown in Fig. 3(a) where the input voltage 1200V and we get the output voltage 190V. Currents through diodes $D1$, $D2$ & $D3$, inductor $L1$, $L2$, capacitor CL , resistor RL are shown in Fig. 3(b), 3(c) & 3(d). The results verify the analysis of the steady state operation. When switch is ON, the current conducts the dual inductors ($L1$ & $L2$) through capacitor (CL) and diode ($D4$). When switch off, the inductor $L1$ supplies the storage energy through the capacitor (CL) and diode ($D1$). Then we get the output voltage 190V and efficiency 98.279%.

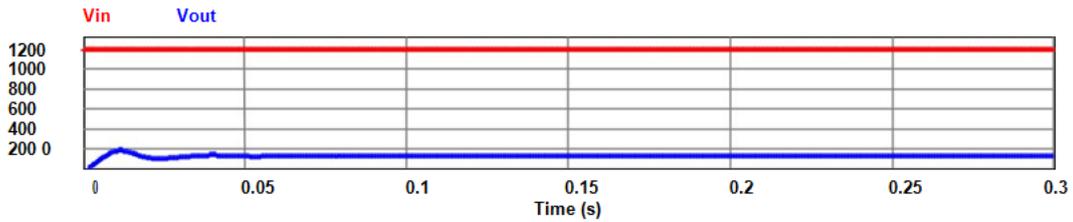


Fig. 3(a). Simulated Input and Output voltage of the proposed DC-DC Buck Converter

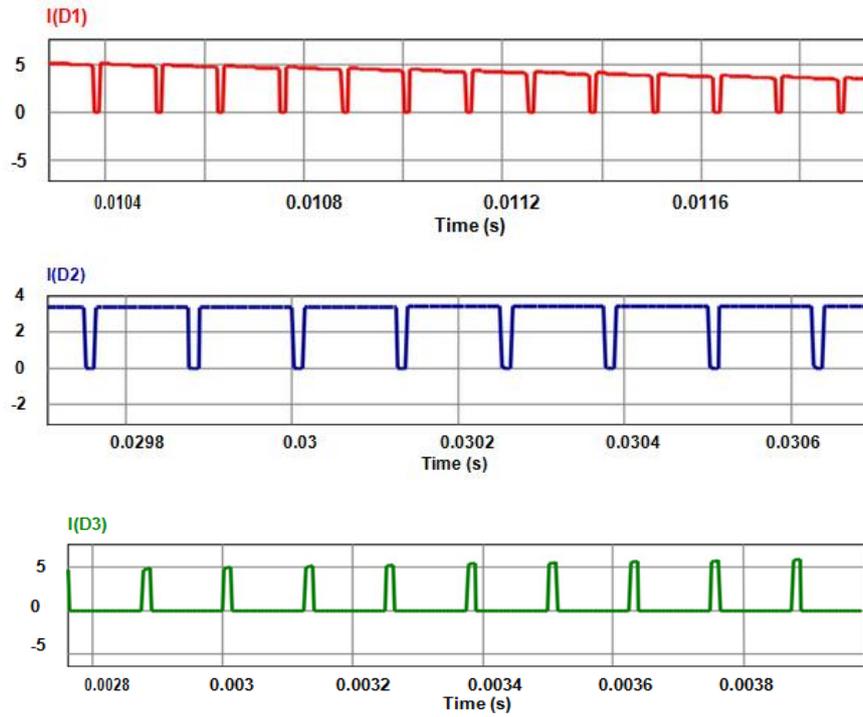


Fig. 3(b). Currents through diodes $D1$, $D2$ & $D3$

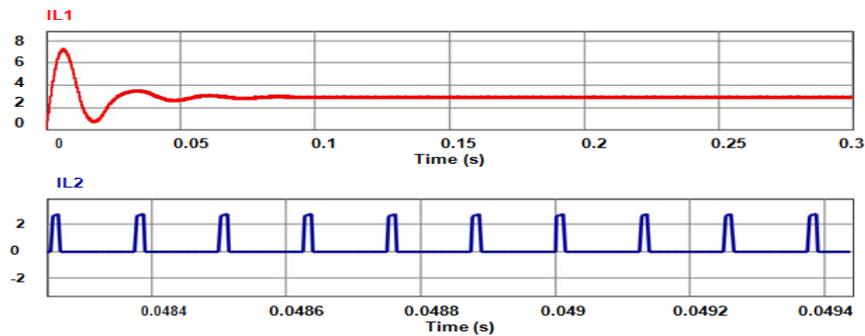


Fig. 3(c). Currents through inductors $L1$, $L2$

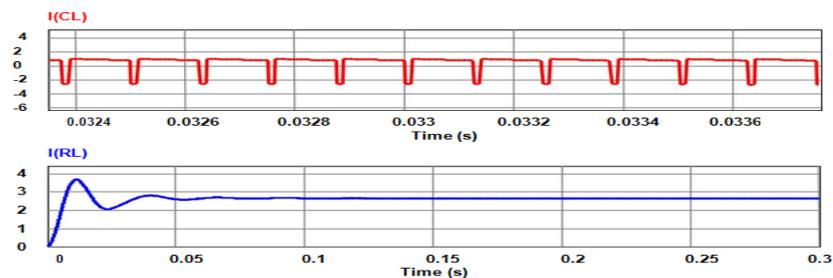


Fig. 3(d). Currents through capacitor CL & resistor RL

B. Quantitative Comparison:

Table II shows the comparative analysis of conventional circuit with the proposed circuit for switching frequency and load variation. Both the circuits are switched at 2KHz-20KHz. It depicts that efficiency of the proposed step-down converter is much better for all the range of efficiency than conventional buck converter. In Table II the circuit performance under load variation is justified for both conventional and Buck based converter for the switching frequency of 8KHz. It also depicts that efficiency of the proposed buck based converter is much better for all the range of efficiency than conventional buck converter. The bar chart shown in Fig. 4 (a) & 4(b) indicates that the comparison of conventional and proposed circuit Efficiency verses switching frequency & Efficiency verses load. Fig. 5 indicates the ratio between duty cycle and voltage gain for the proposed circuit.

Table 2 Performance Comparison Under Switching Frequency And Load Variation

Switching Frequency	Efficiency (%)		Load	Efficiency (%)	
	Conventional Circuit	Proposed Circuit		Conventional Circuit	Proposed Circuit
2K	97.199	97.885	10Ω	92.746	96.601
4K	97.169	97.787	20Ω	95.656	97.606
6K	97.095	97.884	30Ω	96.383	97.560
8K	97.825	98.279	40Ω	96.543	97.294
10K	96.900	97.634	50Ω	96.475	96.931
12K	97.141	97.717	60Ω	96.293	96.516
14K	96.866	97.566	70Ω	96.049	96.125
16K	96.381	96.743	80Ω	95.782	95.732
18K	96.879	97.644	90Ω	95.494	95.334
20K	96.475	96.923	100Ω	95.195	94.935

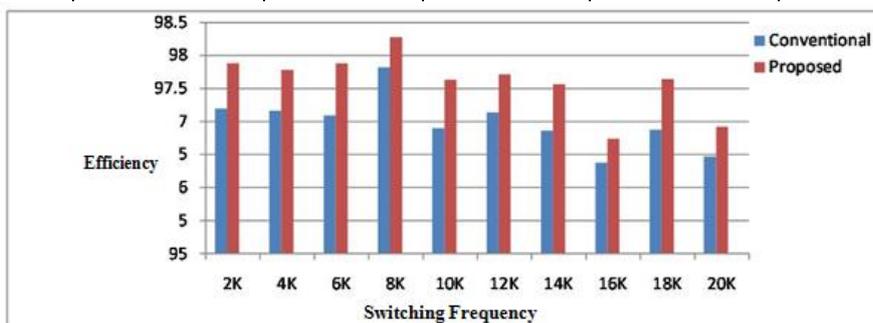


Fig. 4(a). Comparison of conventional and proposed circuit, Efficiency verses switching frequency

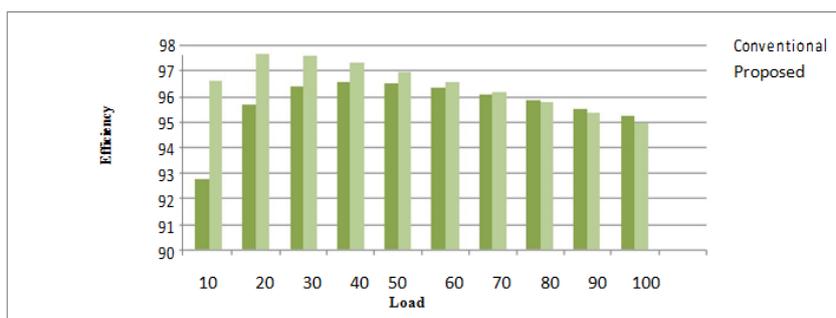
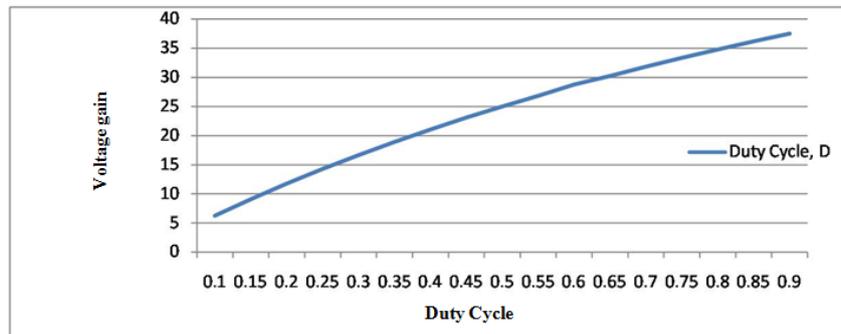


Fig. 4(b). Comparison of conventional and proposed circuit, Efficiency verses load**Fig. 5.** Voltage gain of the proposed circuit throughout the duty cycle 0.1 to 0.9

V. Conclusion

The proposed DC-DC Buck Converter presents low output voltage with high efficiency by using dual inductor and switched capacitor. The energy stored in the inductor (LI) is recycled to improve the performance of the presented DC-DC Buck Converter. Furthermore, the loss of diode and power is reduced. The steady state operation of the proposed DC-DC Buck Converters has been analyzed in details. The results prove of feasibility of the presented DC-DC Buck Converter.

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